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Linearization techniques for band structure calculations in absorbing photonic crystals

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Band structure calculations for photonic crystals require the numerical solution of eigenvalue problems. In this paper, we consider crystals composed of lossy materials with frequency-dependent permittivities. Often, these frequency dependencies are modeled by rational functions, such as the Lorentz model, in which case the eigenvalue problems are rational in the eigenvalue parameter. After spatial discretization using an interior penalty discontinuous Galerkin method, we employ a recently developed linearization technique to deal with the resulting rational matrix eigenvalue problems. In particular, the efficient implementation of Krylov subspace methods for solving the linearized eigenvalue problems is investigated in detail. Numerical experiments demonstrate that our new approach is considerably cheaper in terms of memory and computing time requirements compared to the naive approach of turning the rational eigenvalue problem into a polynomial eigenvalue problem and applying standard linearization techniques.

1 Introduction

This paper is concerned with wave propagation in periodic absorbing media with frequency-dependent material properties. In particular, we consider electromagnetic wave propagation in two-dimensional photonic crystals. In lossless materials, the propagating waves are determined by the spectrum of an underlying PDE eigenvalue problem. Numerical methods for computing this spectrum are well established for frequency-independent materials [10, 3, 7, 29]. Only rather recently, complex dispersion relations $\omega_n(\vec{k})$ in absorbing media with frequency-dependent material properties have been investigated numerically [27, 24, 15, 22, 9]. The frequency-dependent case gives rise to nonlinear eigenvalue problems, for which the development of robust numerical methods is a subtle issue.

In our setting, we assume that the waves propagate in a nonmagnetic material with permittivity ϵ depending on the first two spatial coordinates x_1, x_2 as well as on the frequency ω , but not depending on the third coordinate x_3 . As usual, the electromagnetic wave (E, H) is decomposed into transverse electric (TE) polarized waves $(E_1, E_2, 0, 0, 0, H_3)$ and transverse

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magnetic (TM) polarized waves $(0, 0, E_3, H_1, H_2, 0)$ [17]. This decomposition reduces the full 3D Maxwell equations to scalar 2D Helmholtz equations in H_3 and E_3 , respectively. For simplicity, the discussion will focus on TM polarized waves, but the developed numerical methods can be applied to the TE case as well. By a Bloch ansatz, the original problem on \mathbb{R}^2 is turned into a family of eigenvalue problems in ω posed on the unit cell.

For the finite element discretization of these eigenvalue problems, we use a discontinuous Galerkin framework with curved high-order finite elements. This leads to finite-dimensional eigenvalue problems, which are nonlinear in the eigenvalue parameter ω due to the frequency-dependency of ϵ . The nature of this nonlinearity depends on how the material properties are modeled. Solid materials are typically represented by a sum of Lorentz terms [17], which leads to rational nonlinearities.

Compared to linear eigenvalue problems, the robust numerical solution of nonlinear eigenvalue problems is much more difficult. Newton-type methods as proposed in [4, 30, 31, 19] have excellent local convergence properties but bear the danger that eigenvalues in a specified region of interest could be missed. In special cases, this lack of robustness can be addressed by employing a minmax characterization [35, 36, 37]. Unfortunately, such a characterization does not exist for lossy materials. In the case of rational eigenvalue problems, the most straightforward approach is to multiply by the common denominator and apply standard linearization techniques to the resulting polynomial eigenvalue problem [12, 23]. However, with an increasing number of poles, the degree of the intermediate polynomial eigenvalue problem quickly becomes large, leading to a severe magnification of the problem size during the linearization process. In this paper, an alternative linearization based directly on the rational formulation is proposed along the lines of recent work by Bai and Su [32]. This approach leads to significantly smaller linearizations, especially when the coefficient matrices associated with the rational terms are of low rank. Combined with structure-exploiting Krylov subspace techniques, we demonstrate that this linearization leads to an efficient and robust numerical method for band structure calculations.

The rest of this paper is organized as follows. In Section 2, we describe the mathematical setting of complex band structure calculation and its spatial discretization by an interior penalty discontinuous Galerkin method. Section 3 discusses two different linearization techniques for addressing the resulting rational matrix eigenvalue problem. As the linearized eigenvalue problem is solved by an inverse Arnoldi method, Section 4 is devoted to the efficient numerical inversion for our newly proposed linearization. Finally, Section 5 contains a number of numerical experiments, which demonstrate that the approach proposed in this paper leads to improved performance compared to the standard approach.

2 Band structures of photonic crystals

Non-magnetic two-dimensional photonic crystals are characterized by a periodic frequency-dependent permittivity ϵ . By scaling the lattice constants to unit length, the lattice Γ becomes \mathbb{Z}^2 with the unit cell $\Omega = (0, 1]^2$. In the TM case, the third coordinate E_3 of the electric field satisfies the equation

$$-\Delta E_3(\vec{x}) = \frac{\omega^2}{c^2} \epsilon(\vec{x}, \omega) E_3(\vec{x}), \quad \vec{x} = (x_1, x_2), \quad (1)$$

where ω is the time frequency and c is the speed of light in vacuum. The main tool to study spectral properties of PDEs with periodic coefficients is Floquet-Bloch theory [25, 26, 11, 20].

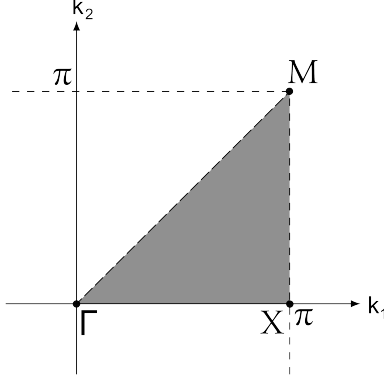


Figure 1: The irreducible Brillouin zone with the symmetry points Γ , X , and M .

A Bloch solution to (1) is a nonzero function of the form

$$E_3(\vec{x}) = e^{i\vec{k}\cdot\vec{x}}u(\vec{x}), \quad (2)$$

where u is a periodic function on Γ and $\vec{k} \in \Omega^* = (-\pi, \pi]^2$ denotes the wave vector. By inserting the Bloch ansatz (2), the original problem (1) posed on the infinite domain \mathbb{R}^2 becomes

$$-(\nabla + i\vec{k}) \cdot (\nabla + i\vec{k})u = \frac{\omega^2}{c^2}\epsilon(\vec{x}, \omega)u, \quad (3)$$

which is a family of eigenvalue problems in ω on the unit cell Ω with periodic boundary conditions. When calculating the dispersion relations $\omega(\vec{k})$ numerically, frequently only a selection of wave vectors \vec{k} along the line segments between the points Γ , X , and M , as shown in Figure 1, is considered. The triangular path formed by these points is called the boundary of the irreducible Brillouin zone [18].

We assume that electromagnetic energy may be transferred into the material, but energy is not transferred from the material into the electromagnetic field. In other words, the material is assumed to be passive [5, 17], which corresponds to the condition

$$\omega\epsilon(\omega) \in \mathbb{C}_+ = \{z \in \mathbb{C} : 0 \leq \arg z < \pi, z \neq 0\} \quad \forall \omega \in \mathbb{C}_+. \quad (4)$$

An important consequence of (4) is $\Im\epsilon(\omega) \geq 0$ for $\omega > 0$. The spectral problem (3) with a passive material model was analyzed in [8], where it was proved that the spectrum is discrete. In principle, any general space-dependent permittivity can be handled when the eigenvalue problem is discretized with finite elements. However, for simplicity, we restrict the permittivity to be piecewise constant with respect to a finite domain partitioning $\Omega = \Omega_1 \cup \dots \cup \Omega_K$:

$$\epsilon(x, \omega) = \sum_{k=1}^K \epsilon_k(\omega)I_k(x), \quad (5)$$

where I_k is the indicator function for the subdomain Ω_k . In general, the permittivities ϵ_n are complicated unknown functions in the frequency ω . It is therefore common to model this

frequency dependency by fitting a rational function satisfying (4) to measurement data. The resulting Lorentz model for solid materials takes the form

$$\epsilon_k(\omega) = \alpha_k + \sum_{\ell=1}^{L_k} \frac{\xi_{k,\ell}^2}{\eta_{k,\ell}^2 - \omega^2 - i\gamma_{k,\ell}\omega}. \quad (6)$$

2.1 Discontinuous Galerkin approximation

In the following, we discretize a weak formulation of the PDE eigenvalue problem (3) with a discontinuous finite element method [2, 1, 14]. The p -version of the finite element method uses polynomial basis functions on a fixed (coarse) mesh and achieves convergence by enhancement of the polynomial degree. More specifically, we use the p -version of the symmetric interior penalty method (SIP) in this paper, since the discretization with a high-order method results, for a given accuracy, in comparably small matrices. Moreover, discontinuous Galerkin methods, such as SIP, are more flexible in the choice of basis functions and the mesh design.

Let \mathcal{V}_h^p denote the space of piecewise polynomials of degree p on a triangulation \mathcal{T}_h of the unit cell $\Omega = \cup_{K \in \mathcal{T}_h} K$. Consider two triangles $K_+, K_- \in \mathcal{T}$ with the outward pointing normals \vec{n}_+, \vec{n}_- on the shared edge $\partial K_+ \cap \partial K_-$. The averages $\{\cdot\}$ and jumps $[\cdot]$ of a piecewise smooth function w on $K_+ \cup K_-$ are then defined as usual:

$$\{w\} = \frac{1}{2}(w_+ + w_-), \quad [w] = w_+ \vec{n}_+ + w_- \vec{n}_-, \quad (7)$$

$$\{\nabla w\} = \frac{1}{2}(\nabla w_+ + \nabla w_-), \quad [\nabla w] = \frac{\partial w_+}{\partial \vec{n}_+} + \frac{\partial w_-}{\partial \vec{n}_-}. \quad (8)$$

For $u_h, v_h \in \mathcal{V}_h^p$ we consider the sesquilinear forms

$$\begin{aligned} a_{\vec{k}}(u_h, v_h) &= \int_{\Omega} \nabla u_h \cdot \nabla \bar{v}_h \, d\vec{x} - \int_{\mathcal{E}} \{\nabla u_h\} \cdot [\bar{v}_h] \, d\vec{x} - \int_{\mathcal{E}} \{\nabla v_h\} \cdot [\bar{u}_h] \, d\vec{x} \\ &\quad + \beta \int_{\mathcal{E}} [u_h] \cdot [\bar{v}_h] \, d\vec{x} - 2i \int_{\Omega} \vec{k} \cdot \nabla u_h \bar{v}_h \, d\vec{x} + \int_{\Omega} \vec{k} \cdot \vec{k} u_h \bar{v}_h \, d\vec{x}, \end{aligned} \quad (9)$$

with \mathcal{E} denoting the set of all edges in \mathcal{T}_h , and

$$b_{\omega}(u_h, v_h) = \int_{\Omega} \epsilon(\vec{x}, \omega) u_h \bar{v}_h \, d\vec{x}. \quad (10)$$

Note that the periodic boundary conditions are imposed weakly by identifying opposite sides of the unit cell and enforcing periodicity of the solution via the corresponding penalty terms in (9); see [34, 9].

The discretized nonlinear eigenvalue problem can now be stated as follows: Find $\omega \in \mathbb{C}$ and a nonzero $u_h \in \mathcal{V}_h^p$ such that

$$a_{\vec{k}}(u_h, v_h) = \omega^2 b_{\omega}(u_h, v_h), \quad (11)$$

holds for all $v_h \in \mathcal{V}_h^p$. A numerical method based on (11) is symmetric and stable provided that the penalty parameter β is sufficiently large [1, 34]. Based on the experiments from [9], we use

$$\beta = 20 \frac{(p+1)(p+2)}{2} \frac{|\partial K|}{|K|}, \quad (12)$$

where $|\partial K|$ denotes the length of the edge and $|K|$ designates the area of the smaller of the adjacent triangles. The global approximation u is computed as a direct sum of local solutions u^K at every triangle K in the mesh.

To generate the mesh, we use the software package *Emc²* [28]. Parts of the geometry are frequently curved, for example, a circular interface between an inclusion and the matrix. In order to obtain high accuracy by increasing the polynomial degree p of the finite elements, it is therefore mandatory to use curved triangles combined with a Gordon-Hall blending [16, 14]. To assemble (11), we build on the MATLAB version [14] of NUDG++ (www.nudg.org). The local matrices on a triangle K are assembled into global matrices and, finally, the discretized nonlinear eigenvalue problem takes the form

$$Gu - \omega^2 \sum_{k=1}^K \epsilon_k(\omega) M_k u = 0, \quad (13)$$

where u is now a vector containing the coefficients of u_h . The number of degrees of freedom (N) is $\frac{1}{2}(p+1)(p+2)$ times the number of triangles in the mesh. The finite element discretization error for this formulation was addressed in [9]. The focus of this paper is on efficient algorithms for solving the discretized problem (13).

3 Linearization of rational eigenvalue problems

In this section, we discuss the numerical solution of the finite-dimensional nonlinear eigenvalue problem (13). For the moment, we consider a two-component model ($K = 2$) with one of the materials having constant permittivity α_1 . For the other material, the permittivity is described by the Lorentz model (6), resulting in the rational eigenvalue problem

$$Gu - \omega^2 \alpha_1 M_1 u - \omega^2 \left(\alpha_2 + \sum_{\ell=1}^L \frac{\xi_\ell^2}{\eta_\ell^2 - \omega^2 - i\gamma_\ell \omega} \right) M_2 u = 0. \quad (14)$$

Later on, in Section 3.3, we will comment on the extension to arbitrarily many components.

3.1 Linearization via a polynomial eigenvalue problem

Simply through multiplication by the denominators, the rational eigenvalue problem (14) can be turned into a polynomial eigenvalue problem. More specifically, we obtain

$$\left[\prod_{\ell=1}^L (\eta_\ell^2 - \omega^2 - i\gamma_\ell \omega) \right] (G - \omega^2 (\alpha_1 M_1 + \alpha_2 M_2)) u - \omega^2 \left[\sum_{\ell=1}^L \xi_\ell^2 \prod_{m \neq \ell} (\eta_m^2 - \omega^2 - i\gamma_m \omega) \right] M_2 u = 0. \quad (15)$$

The eigenvalues of (15) coincide with the eigenvalues of (14), provided that ω is bounded away from the poles. Sorting coefficients, the matrix polynomial (15) can be written in the form

$$\omega^{2L+2} P_{2L+2} u + \omega^{2L+1} P_{2L+1} u + \dots + P_0 u = 0, \quad (16)$$

for certain matrices P_0, \dots, P_{2L+2} . The usual way to solve (16) is to apply a standard eigenvalue solver to an equivalent linear eigenvalue problem

$$\mathcal{L}_1 u = \omega \mathcal{L}_2 u. \quad (17)$$

This linearization is not unique [23]; a popular choice is the companion linearization

$$\mathcal{L}_1 = \begin{bmatrix} & I & & \\ & & \ddots & \\ & & & I \\ -P_0 & -P_1 & \cdots & -P_{2L+1} \end{bmatrix}, \quad \mathcal{L}_2 = \begin{bmatrix} I & & & \\ & \ddots & & \\ & & I & \\ & & & P_{2L+2} \end{bmatrix}. \quad (18)$$

An obvious disadvantage of this approach is the increase of the dimension from N to $(2L+2) \cdot N$. More subtle disadvantages include a certain loss of sparsity and numerical instability due to numerical cancellation during the process of forming P_j .

3.2 Direct linearization

To avoid the above-mentioned disadvantages, we consider a direct linearization of (14) without the intermediate transformation to a matrix polynomial, based on the approach proposed by Bai and Su [32]. To apply this approach, we first bring the rational terms in (14) to proper form:

$$\frac{\xi_\ell^2 \omega^2}{\eta_\ell^2 - \omega^2 - i\gamma_\ell \omega} = -\xi_\ell^2 + \frac{\xi_\ell^2(\eta_\ell^2 - i\gamma_\ell \omega)}{\eta_\ell^2 - \omega^2 - i\gamma_\ell \omega}.$$

The second term in this sum can be represented as $b_\ell^T (A_\ell - \omega E_\ell)^{-1} b_\ell$, where

$$A_\ell = \frac{2\theta_\ell}{\xi_\ell^2} \begin{bmatrix} \theta_\ell - \frac{1}{2}i\gamma_\ell & 0 \\ 0 & \theta_\ell + \frac{1}{2}i\gamma_\ell \end{bmatrix}, \quad E_\ell = \frac{2\theta_\ell}{\xi_\ell^2} \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}, \quad b_\ell = \begin{bmatrix} \theta_\ell - \frac{1}{2}i\gamma_\ell \\ \theta_\ell + \frac{1}{2}i\gamma_\ell \end{bmatrix}, \quad (19)$$

with $\theta_\ell = \sqrt{\eta_\ell^2 - \frac{1}{4}\gamma_\ell^2}$. Setting $A = \text{diag}(A_1, \dots, A_L)$, $E = \text{diag}(E_1, \dots, E_L)$, and $b = [b_1^T, \dots, b_L^T]^T$, the eigenvalue problem (14) becomes

$$(\hat{G} - \omega^2 \hat{M} - b^T (A - \omega E)^{-1} b M_2) u = 0, \quad (20)$$

where

$$\hat{G} = G + \sum_{\ell=1}^L \xi_\ell^2 M_2, \quad \hat{M} = \alpha_1 M_1 + \alpha_2 M_2.$$

To proceed from (20), we note that M_2 is the mass matrix for the FE basis functions supported on the frequency-dependent subdomain Ω_2 . By a suitable ordering of the nodes, M_2 takes the form $M_2 = \text{diag}(0, \check{M}_2)$ with a symmetric positive definite matrix $\check{M}_2 \in \mathbb{R}^{N_2 \times N_2}$. Note that $N_2 < N$ corresponds to the number of basis functions supported on Ω_2 . Performing a Cholesky factorization of \check{M}_2 , we can therefore write $M_2 = F^T F$ for some $F \in \mathbb{R}^{N_2 \times N}$. This gives

$$\begin{aligned} b^T (A - \omega E)^{-1} b M_2 &= b^T (A - \omega E)^{-1} b \otimes F^T F \\ &= \mathcal{B}^T (\mathcal{A} - \omega \mathcal{E})^{-1} \mathcal{B}, \end{aligned}$$

where $\mathcal{B} := b \otimes F$, $\mathcal{A} := A \otimes I_{N_2}$, $\mathcal{E} := E \otimes I_{N_2}$. Note that \otimes denotes the Kronecker product of two matrices; see [13] for details.

Combined with (20),

$$(\hat{G} - \omega^2 \hat{M} - \mathcal{B}^T (\mathcal{A} - \omega \mathcal{E})^{-1} \mathcal{B}) u = 0, \quad (21)$$

the eigenvalue problem can now be linearized by introducing the auxiliary vectors $v = -(\mathcal{A} - \omega\mathcal{E})^{-1}\mathcal{B}u$ and $w = \omega u$:

$$\begin{bmatrix} \hat{G} & \mathcal{B}^T & 0 \\ \mathcal{B} & \mathcal{A} & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} u \\ v \\ w \end{bmatrix} = \omega \begin{bmatrix} 0 & 0 & \hat{M} \\ 0 & \mathcal{E} & 0 \\ I & 0 & 0 \end{bmatrix} \begin{bmatrix} u \\ v \\ w \end{bmatrix}. \quad (22)$$

This is a generalized linear eigenvalue problem of order $2LN_2 + 2N$, which compares favorably with the order $2LN + 2N$ of the linear eigenvalue problem.

3.3 Extension to $K > 2$ materials

The derivation above can be directly extended to K subdomains Ω_k , $k = 1, \dots, K$, for which the frequency dependency is described by the parameters $\alpha_k, \xi_{k,\ell}, \eta_{k,\ell}, \gamma_{k,\ell}$ in the Lorentz model (6). Based on these parameters, matrices $A_{k,\ell}, E_{k,\ell}$, and vectors $b_{k,\ell}$, $k = 1, \dots, K$, $\ell = 1, \dots, L_k$, are defined analogously as in (19). After factorizing the mass matrices $M_k = F_k^T F_k$, $F_k \in \mathbb{R}^{N_k \times N}$ as before and defining

$$\begin{aligned} \hat{G} &= G + \sum_{k=1}^K \sum_{\ell=1}^{L_k} \xi_{k,\ell}^2 M_k, \\ \hat{M} &= \alpha_1 M_1 + \dots + \alpha_K M_K, \\ \mathcal{A} &= \text{diag}(A_1 \otimes I_{N_1}, \dots, A_K \otimes I_{N_K}), \\ \mathcal{E} &= \text{diag}(E_1 \otimes I_{N_1}, \dots, E_K \otimes I_{N_K}), \\ \mathcal{B} &= [(b_1 \otimes F_1)^T, \dots, (b_K \otimes F_K)^T]^T, \end{aligned}$$

where

$$\begin{aligned} A_k &= \text{diag}(A_{k,1}, \dots, A_{k,L_k}), \\ E_k &= \text{diag}(E_{k,1}, \dots, E_{k,L_k}), \\ b_k &= [b_{k,1}^T, \dots, b_{k,L_k}^T]^T, \end{aligned}$$

for $k = 1, \dots, K$, we again obtain a linearized eigenvalue problem of precisely the form (22).

4 Solution of the linearized eigenvalue problem

Both linearizations discussed in Section 3 lead to a linear eigenvalue problem of the form $\mathcal{L}_1 u = \omega \mathcal{L}_2 u$. In our application, we aim at computing a few eigenvalues of *smallest* magnitude (close to the real axis). Premultiplying by \mathcal{L}_1^{-1} and dividing by ω leads to

$$\mathcal{L}_1^{-1} \mathcal{L}_2 u = \omega^{-1} u. \quad (23)$$

Consequently, ω is an eigenvalue for $\mathcal{L}_1 u = \omega \mathcal{L}_2 u$ if and only if ω^{-1} is an eigenvalue of the matrix $\Phi := \mathcal{L}_1^{-1} \mathcal{L}_2$. Thus, our aim becomes the computation of eigenvalues of *largest* magnitude for Φ , a task which can be easily addressed using standard Krylov subspace methods, such as the implicitly restarted Arnoldi method [21].

These Krylov subspace methods are based on repeated matrix-vector multiplication with Φ , which, in particular, requires the efficient inversion of \mathcal{L}_1 . For the companion linearization (18),

any sparse direct solver automatically takes advantage of the sparsity structure of \mathcal{L}_1 . For the direct linearization (22), it is not obvious how the structure of \mathcal{L}_1 can be exploited and the rest of this section will be devoted to this issue.

First, we recall that for the direct linearization (22), the reciprocal eigenvalue problem (23) takes the form

$$\underbrace{\begin{bmatrix} \begin{bmatrix} \hat{G} & \mathcal{B}^T \\ \mathcal{B} & \mathcal{A} \end{bmatrix}^{-1} & \\ & I \end{bmatrix}}_{=\Phi} \begin{bmatrix} 0 & 0 & \hat{M} \\ 0 & \mathcal{E} & 0 \\ I & 0 & 0 \end{bmatrix} \begin{bmatrix} u \\ v \\ w \end{bmatrix} = \frac{1}{\omega} \begin{bmatrix} u \\ v \\ w \end{bmatrix}. \quad (24)$$

We then factorize

$$\begin{bmatrix} \hat{G} & \mathcal{B}^T \\ \mathcal{B} & \mathcal{A} \end{bmatrix} = \begin{bmatrix} I & \mathcal{B}^T \mathcal{A}^{-1} \\ 0 & I \end{bmatrix} \begin{bmatrix} \hat{G} - \mathcal{B}^T \mathcal{A}^{-1} \mathcal{B} & 0 \\ 0 & \mathcal{A} \end{bmatrix} \begin{bmatrix} I & 0 \\ \mathcal{A}^{-1} \mathcal{B} & I \end{bmatrix}. \quad (25)$$

From the definitions of \hat{G} , \mathcal{A} , and \mathcal{B} in Section 3.2, we find that

$$\hat{G} - \mathcal{B}^T \mathcal{A}^{-1} \mathcal{B} = \hat{G} - b^T A b \otimes F^T F = \hat{G} - \sum_{\ell=1}^L b_\ell^T A_\ell^{-1} b_\ell \otimes M_2 = \hat{G} - \sum_{\ell=1}^L \xi_\ell^2 M_2 = G.$$

Inserting this relation into (25) and inverting yields

$$\begin{aligned} \begin{bmatrix} \hat{G} & \mathcal{B}^T \\ \mathcal{B} & \mathcal{A} \end{bmatrix}^{-1} &= \begin{bmatrix} I & 0 \\ -\mathcal{A}^{-1} \mathcal{B} & I \end{bmatrix} \begin{bmatrix} G^{-1} & 0 \\ 0 & \mathcal{A}^{-1} \end{bmatrix} \begin{bmatrix} I & -\mathcal{B}^T \mathcal{A}^{-1} \\ 0 & I \end{bmatrix} \\ &= \begin{bmatrix} G^{-1} & -G^{-1} \mathcal{B}^T \mathcal{A}^{-1} \\ -\mathcal{A}^{-1} \mathcal{B} G^{-1} & \mathcal{A}^{-1} + \mathcal{A}^{-1} \mathcal{B} G^{-1} \mathcal{B}^T \mathcal{A}^{-1} \end{bmatrix}. \end{aligned}$$

Hence, the matrix-vector product $[\tilde{u}^T, \tilde{v}^T, \tilde{w}^T]^T := \Phi [u^T, v^T, w^T]^T$ is given by the formulas

$$\begin{aligned} \tilde{u} &= G^{-1} \hat{M} w - G^{-1} \mathcal{B}^T \mathcal{A}^{-1} \mathcal{E} v = G^{-1} (\hat{M} w - \mathcal{B}^T \mathcal{A}^{-1} \mathcal{E} v) \\ \tilde{v} &= \mathcal{A}^{-1} \mathcal{E} v + \mathcal{A}^{-1} \mathcal{B} G^{-1} \mathcal{B}^T \mathcal{A}^{-1} \mathcal{E} v - \mathcal{A}^{-1} \mathcal{B} G^{-1} \hat{M} w = \mathcal{A}^{-1} \mathcal{E} v - \mathcal{A}^{-1} \mathcal{B} \tilde{u} \\ \tilde{w} &= u. \end{aligned}$$

Consequently, the major computational work when applying Φ consists in solving a linear system with the stiffness matrix G . All the remaining operations are comparatively cheap:

- a matrix-vector product with the sparse matrix \hat{M} ,
- a matrix-vector product with

$$\mathcal{A}^{-1} \mathcal{E} = A^{-1} E \otimes I_{N_2} = \text{diag}(A_1^{-1} E_1, \dots, A_L^{-1} E_L) \otimes I_{N_2},$$

which is just a diagonal scaling, and



Figure 2: Unit cell of the benchmark setup

- matrix-vector products with

$$\mathcal{B}^T \mathcal{A}^{-1} \mathcal{E} = b^T A^{-1} E \otimes F^T = \begin{bmatrix} 1 \\ -1 \\ \vdots \\ 1 \\ -1 \end{bmatrix} \otimes F^T = \begin{bmatrix} 1 \\ \vdots \\ 1 \end{bmatrix} \otimes \begin{bmatrix} 1 \\ -1 \end{bmatrix} \otimes F^T,$$

$$\mathcal{A}^{-1} \mathcal{B} = A^{-1} b \otimes F = \begin{bmatrix} \frac{\xi_1^2}{2\theta_1} \\ \frac{\xi_1^2}{2\theta_1} \\ \vdots \\ \frac{\xi_L^2}{2\theta_L} \\ \frac{\xi_L^2}{2\theta_L} \end{bmatrix} \otimes F = \begin{bmatrix} \frac{\xi_1^2}{2\theta_1} \\ \vdots \\ \frac{\xi_L^2}{2\theta_L} \end{bmatrix} \otimes \begin{bmatrix} 1 \\ 1 \end{bmatrix} \otimes F,$$

which are highly structured.

As the stiffness matrix G is Hermitian and – for nonzero wave vectors \vec{k} – also positive definite, we employ a sparse Cholesky decomposition for solving linear systems with G . Unfortunately, G depends on the wave vector and, therefore, needs to be refactorized whenever the wave vector changes. Still, we benefit significantly from reusing the Cholesky factors since the number of wave vectors under consideration is typically much smaller than the total number of linear systems to be solved.

5 Numerical examples

In order to evaluate the numerical method proposed in Section 4, we consider the following setup: A disk of radius $r = 0.2$ consisting of a material with frequency-dependent permittivity is located in the center of the unit cell; see Figure 2. The rest of the unit cell is filled with air (relative permittivity $\epsilon_1 = 1$). The permittivity of the frequency-dependent material is governed by the Lorentz model (6), leading to a rational eigenvalue problem of the form (14).

We may solve this eigenvalue problem via the *standard linearization* based on an intermediate transformation to a polynomial eigenvalue problem (see Section 3.1) or via the direct linearization developed in Section 3.2. The computations have been performed under MATLAB 7.10 on a cluster of 16 AMD Opteron 8356 processors with 64 GB of shared memory. The

ℓ	1	2
ξ_ℓ^2	98.6960	197.3921
η_ℓ^2	55.2698	63.1655
γ_ℓ	0.0063	0.1257
$\alpha = 2.0$		

Table 1: Material parameters for the first set of experiments.

problem size		standard linearization		direct linearization	
p	#dofs	size	comp. time (s)	size	comp. time (s)
2	288	1728	15.2	720	6.4
4	720	4320	47.0	1800	16.7
6	1344	8064	114	3360	46.7
8	2160	12960	270	5400	130
12	4368	26208	980	10920	430
18	9120	54720	4774	22800	1424

Table 2: A comparison of the resulting sizes and computing times between the standard and direct linearizations for a Lorentz model with 2 terms (see Table 1).

ℓ	1	2	3	4	5	6	7
ξ_ℓ^2	416.6166	352.7054	-339.9124	492.5687	-19.6143	-527.5597	98.0101
η_ℓ^2	92.1086	71.6269	71.4552	227.8301	47.4923	93.5605	121.3762
γ_ℓ	2.7820	0.9597	0.9500	13.1508	9.2697	3.2624	2.2712
$\alpha = 1.143$							

Table 3: Material parameters for the second set of experiments, representing 66% porous silicon, scaled with the lattice constant $a = 500$ nm.

problem size		standard linearization		direct linearization	
p	#dofs	size	comp. time (s)	size	comp. time (s)
2	288	4608	—	1584	14.9
4	720	11520	—	3960	35.2
6	1344	21504	—	7392	89.2
8	2160	34560	—	11880	222
12	4368	69888	—	24024	817
18	9120	145920	—	50160	2081

Table 4: A comparison between the sizes of the standard and direct linearizations for a Lorentz model with 7 terms (see Table 3); in the direct case also computing times are given.

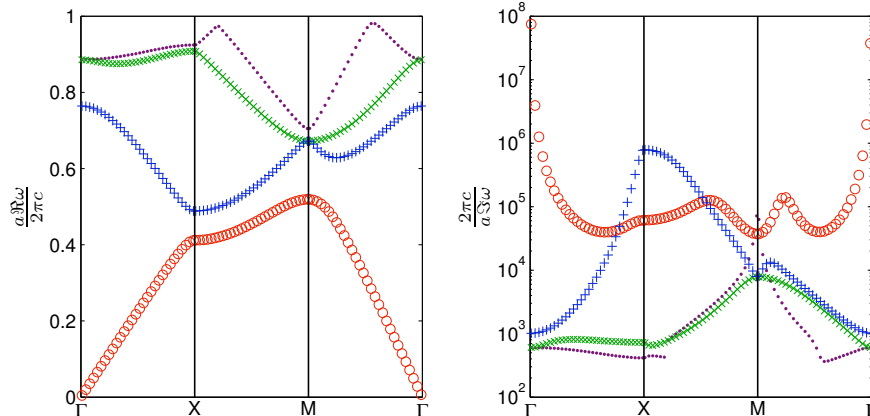


Figure 3: The band diagram computed by our method for the 7-terms Lorentz model of 66% porous silicon (see Table 3).

MATLAB interface to UMFPACK [6] has been utilized to carry out sparse direct factorizations. The linearized eigenvalue problems have been solved by the MATLAB interface `eigs` to ARPACK [21]. All reported computing times have been averaged over 5 runs.

In our first set of experiments, we have used a synthetic 2-term Lorentz model with the coefficients given in Table 1. We have experimented with different levels of p -refinement in the discontinuous Galerkin approximation; see Section 2.1. Table 2 contains the resulting matrix sizes for the standard and direct linearizations along with the computing times for determining the 10 eigenvalues of smallest magnitude for each of 30 different wave vectors distributed along the boundary of the irreducible Brillouin zone. It turns out that our new approach based on direct linearization is faster by a factor between 2.0 and 3.3.

In our second set of experiments, we have used 7 Lorentz terms to model the frequency-dependent permittivity of 66% porous silicon [33, Table II, p. 523], scaled to our lattice constant $a = 500$ nm. The corresponding parameters can be found in Table 3. Again, we have computed the 10 eigenvalues of smallest magnitude for each of 30 different wave vectors distributed along the boundary of the irreducible Brillouin zone. Table 4 summarizes the results for different levels of p -refinement. For the standard linearization, the increased number of Lorentz terms does not only severely complicate the analytic derivation of the polynomial coefficients but also results in excessive memory requirements; therefore, we have only stated the expected size of the linearization but no computing times. For the direct linearization, it is interesting to note that the computing times only grow very moderately (roughly by a factor of 2) when increasing the number of Lorentz terms from 2 to 7. This is a consequence of the high level of structure exploitation in the derivation of our new method. Finally, the resulting dispersion relations are depicted in Figure 3.

6 Conclusions and outlook

We have proposed an algorithm for solving rational eigenvalue problems arising in band structure calculations for photonic crystals containing frequency-dependent materials. By taking the rich structure of these eigenvalue problems into account, we arrive at linearizations of

reduced size, which considerably reduces the computational effort in terms of memory and execution time. Future work includes the efficient handling of non-zero shifts in the linearized eigenvalue solver, which may extend the range of applicability of our method. Moreover, it might be beneficial to combine our method with eigenvalue continuation techniques when repeatedly solving the eigenvalue problem for different but nearby wave vectors, a typical situation in band structure calculations. Unfortunately, the continuation of several eigenvalues for nonlinear eigenvalue problems is presently not very well understood.

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